

Scaling of Vibration Energy Harvester

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Motivation

- Vibrational energy harvesting become viable alternatives because of the limitations of batteries, along with the reduction of power consumption and size. [1]
- Working frequency range of conventional resonance devices can be increased using nonlinearities obtained through geometrical design. [2]
- As devices continue to shrink, it is also necessary to miniaturize energy harvesters while maximizing power.

Background

- Wideband MEMS energy harvesters
 - Well-suited to extract power from a wide spectrum of vibrations.
 - Greater output power and wide bandwidth due to inclined springs. [3]
- Applications:
 - Bridge or structural constructions wireless monitoring sensors.
 - Wearable and implantable sensors.
 - Automotive tire pressure monitoring systems (TPMSs).

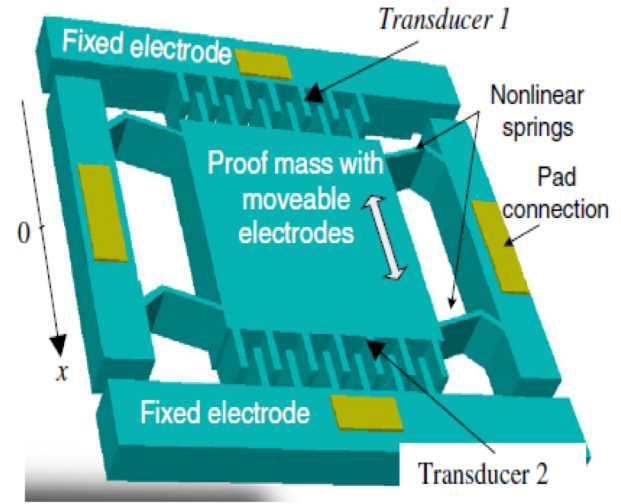
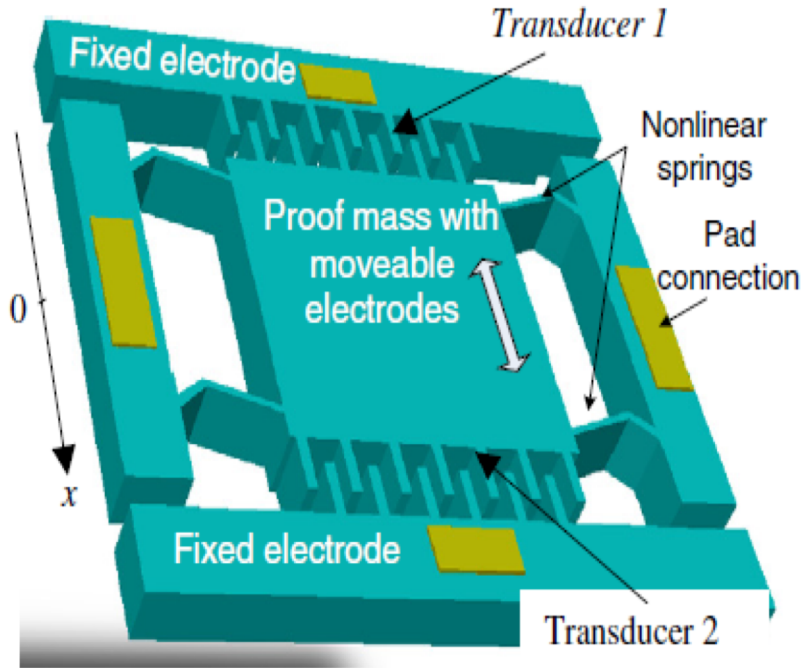


Figure 1. Schematic drawing of a MEMS electrostatic energy harvester with nonlinear springs. [4]

Device Description

- Electrostatic vibrational energy harvester.
- Angled spring to achieve nonlinearities.
- Fabrication: SOI DRIE process with three photolithography masks.

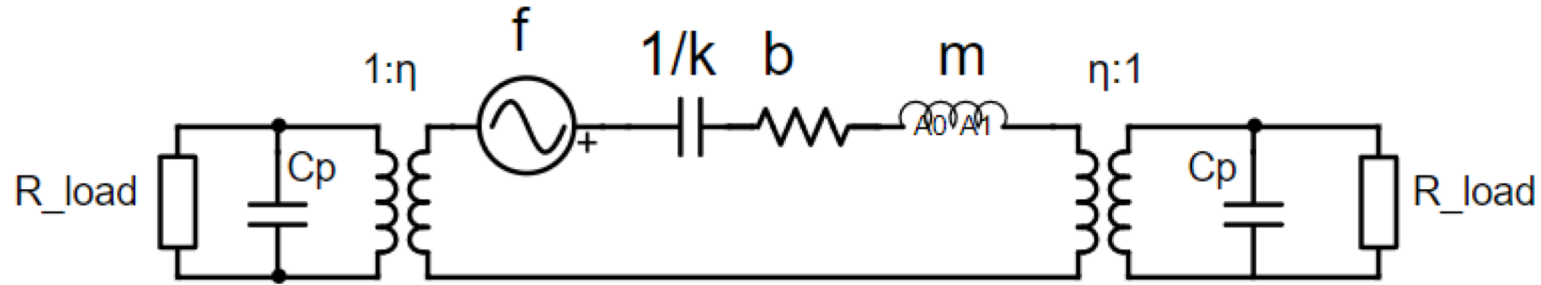


State-of-the-art Vibration Energy Harvesting Devices

Hypothesis: Scaling the vibration energy harvester down will increase power output per mass of the device.

Ref	Year	Specific Power (W/kg)
4	2010	1.22e-2
5	2011	1.27e-3
6	2013	3.10e-5
7	2013	9.35e-3
Our Scaling Target		>1e-1

Modeling the Harvester with Equivalent Circuit



From this we can find the power transferred across both loads as

$$P = 2 \times R_{load} \left[\frac{1}{\sqrt{2}} \cdot \frac{1}{1 + sCR_{load}} \cdot \frac{f\eta}{b + \frac{k}{s} + sm + 2\eta^2 \left(\frac{R_{load}}{1 + sCR_{load}} \right)} \right]^2$$

$$\eta = 2N_f V_p \frac{\epsilon_0 h}{g}$$

$$R = \frac{b}{\sqrt{4\eta^4 + b^2 C^2 \omega_0^2}}$$

Model with Experimental Results

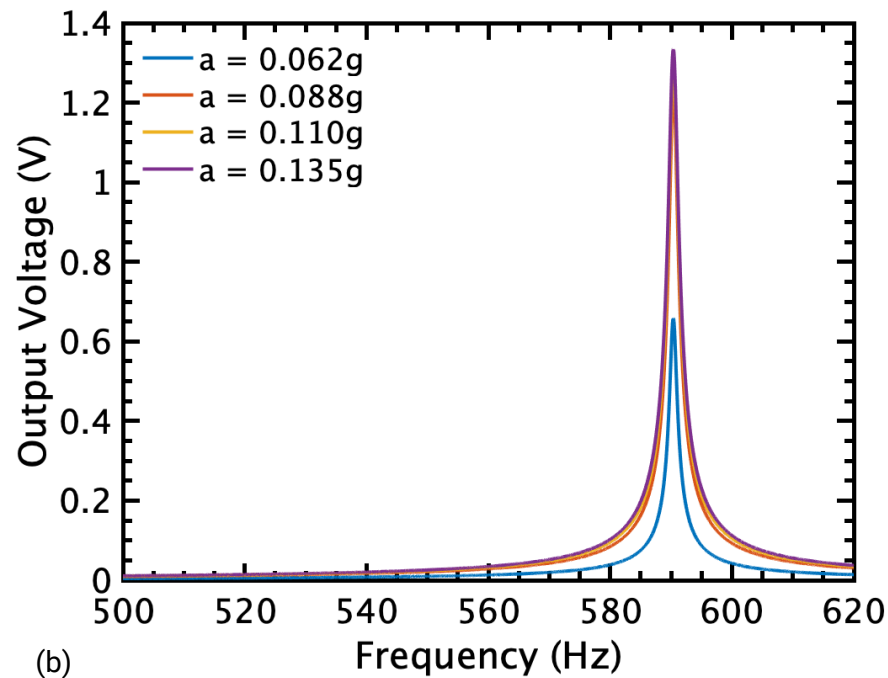
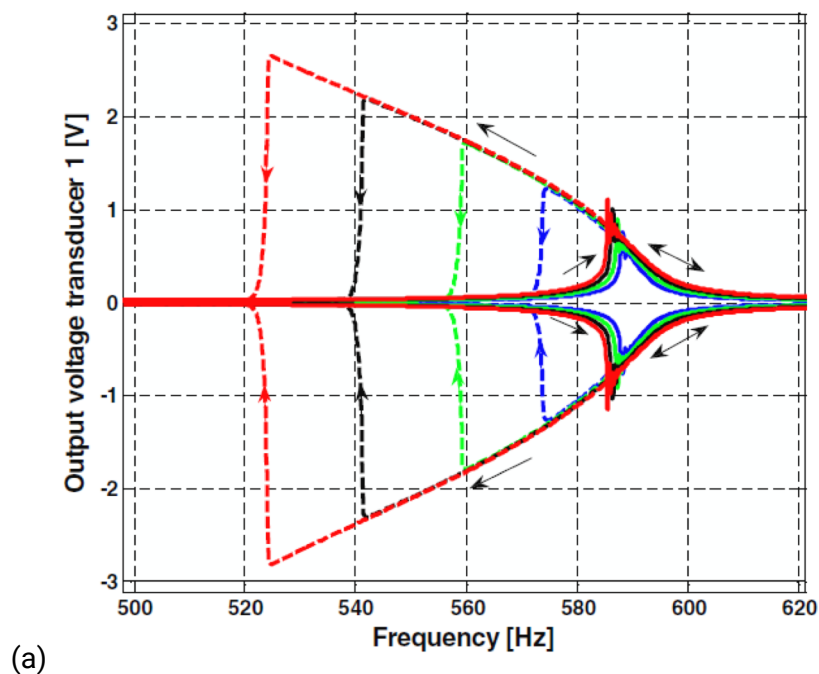


Figure 2: Peak output voltages as a function of frequency (a) Literature experimental results (b) Generated model.

Model with Experimental Results

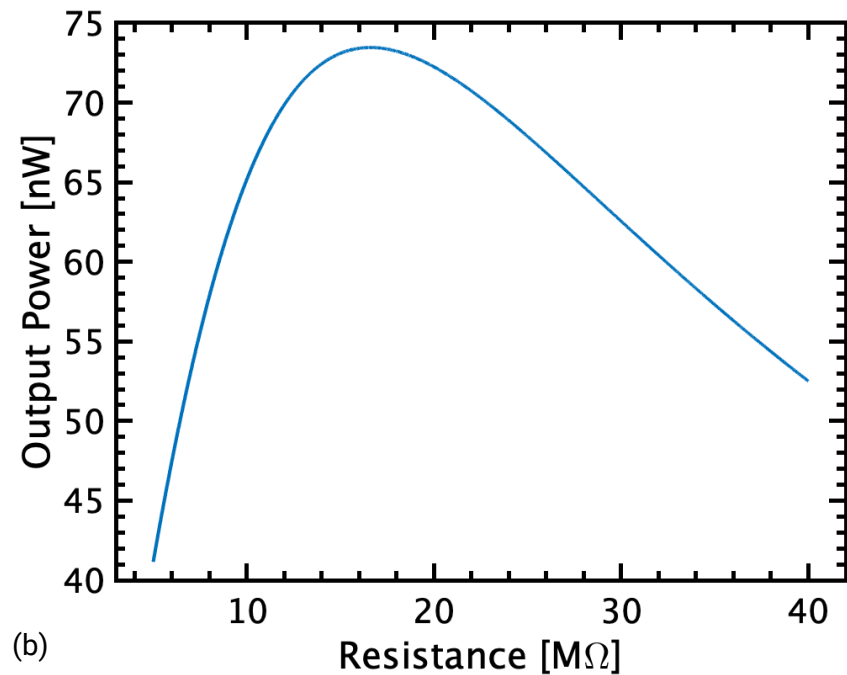
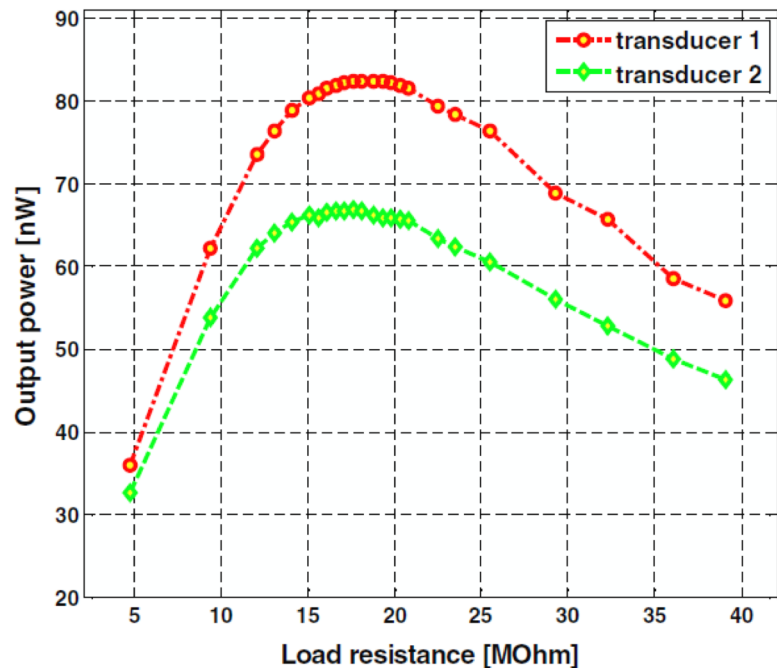


Figure 3: Output power as a function of load resistance for sinusoidal vibration, with acceleration of 0.14g and a bias voltage of 28.4 V (a) Literature experimental results (b) Generated model.

Two Types of Scaling

$$P = 2 \times R_{load} \left[\frac{1}{\sqrt{2}} \cdot \frac{1}{1 + sCR_{load}} \cdot \frac{f\eta}{b + \frac{k}{s} + sm + 2\eta^2 \left(\frac{R_{load}}{1 + sCR_{load}} \right)} \right]^2 \quad \eta = 2N_f V_p \frac{\epsilon_0 h}{g}$$

- Scaling Finger Gaps

- Only affects η
- Equal effect by decreasing the gaps as increasing the number of fingers
- Does not affect quality factor or resonant frequency significantly
- Affects load resistance

- Scaling Entire Structure

- Affects damping, stiffness, mass and gaps
- Reduces allowed bias voltage (changes η)
- Shifts resonant frequency but should not change quality factor
- Affects load resistance

Final Structure Specifications

Description	Original	Gap Scaled (S=7.5)	Fully Scaled (S=1000)
Mass	35.25 mg	35.25 mg	35.25 pg
Thickness	300 μm	300 μm	300 nm
Beam Length	~ 1400 μm	~ 1400 μm	1.4 μm
Beam Width	20 μm	20 μm	20 nm
Transducer Gap	15 μm	2 μm	15 nm
Finger width	15 μm	2 μm	15 nm
Initial Capacitive Overlap	~ 120 μm	~ 120 μm	120 nm
Number of fingers	128	960	128
Bias Voltage	28.4 V	10.1 V	0.1 V

Model Assumptions for Scaling

- Damping Coefficient

- Damping from Couette flow dominates

$$b = \mu A / g$$

- Area scales but undercut of release etch does not
- Viscous damping from the gaps around the fingers is negligible

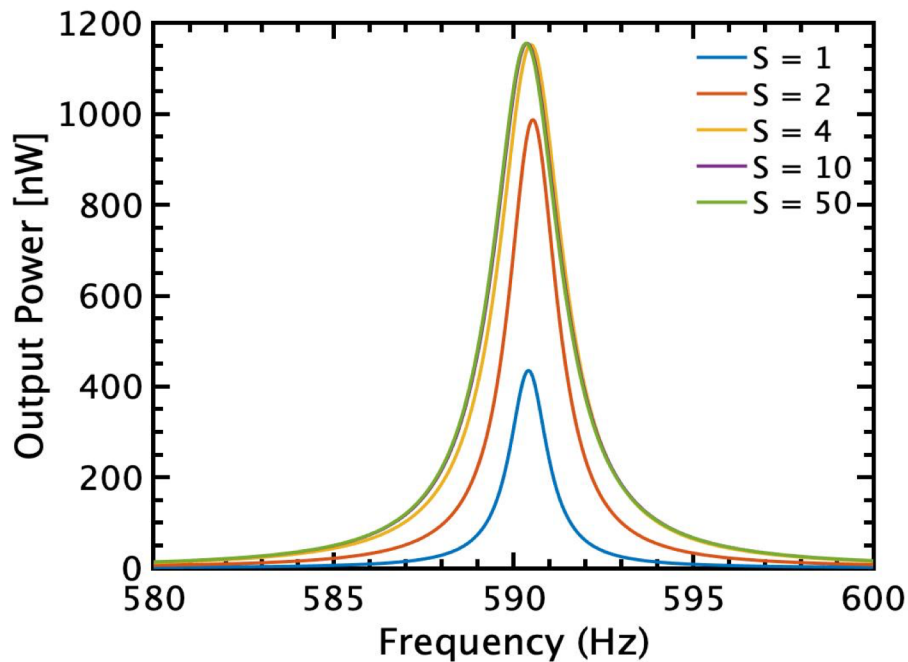
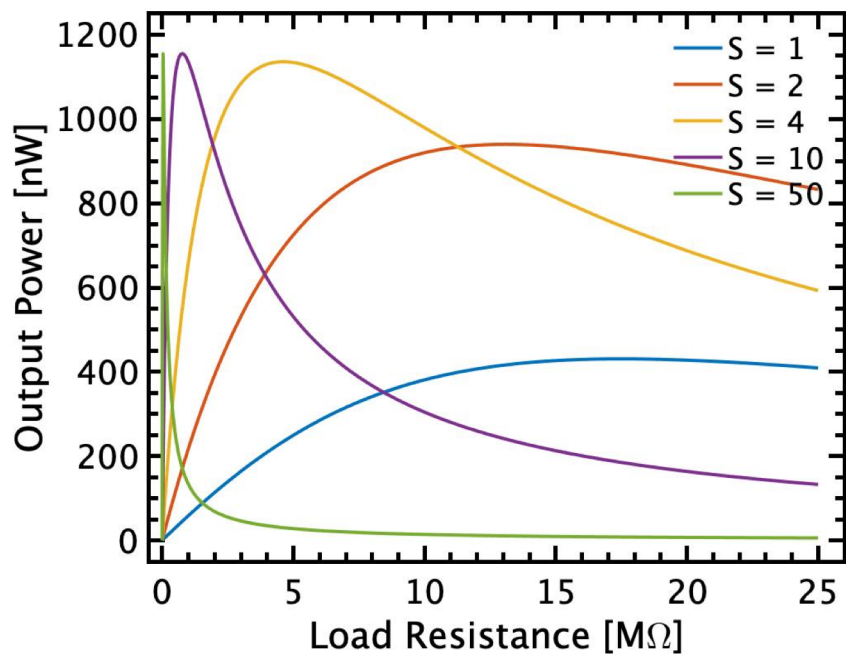
- The spring coefficient is near linear in the region of operation

- k scales with S

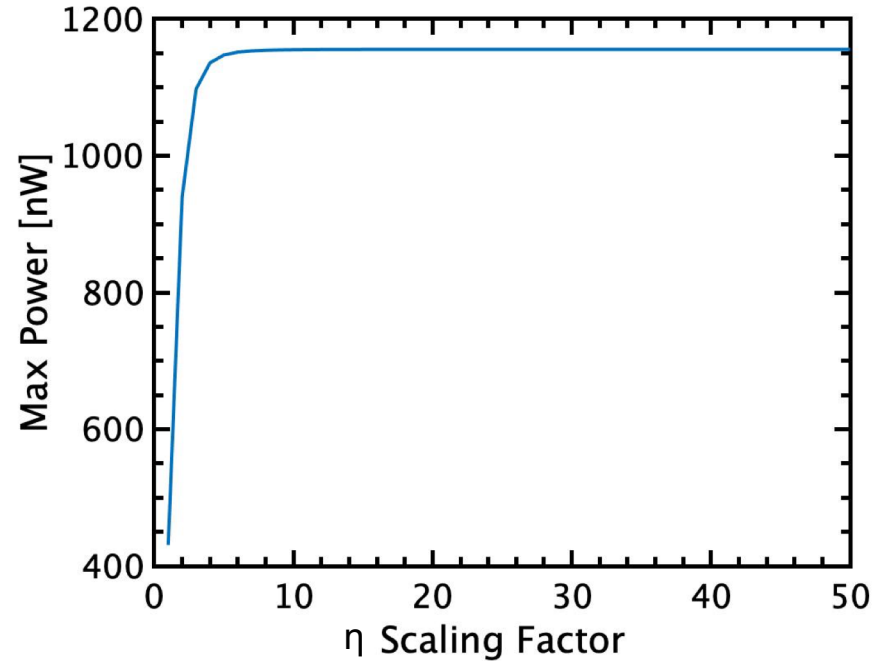
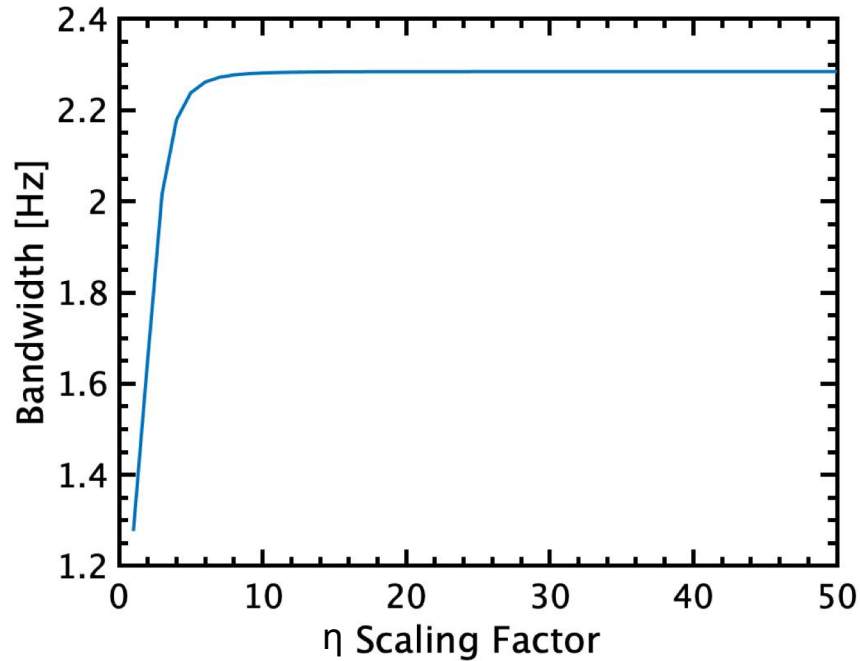
- All length dimensions of capacitance scale with S

- C scales with S

Scaling Finger Gaps



Trend of Scaling Finger Gaps



Interpretation

$$P_{resonance} = R \left(\frac{1}{1 + sCR} \cdot \frac{f\eta}{b + 2\eta^2 \left(\frac{R}{1+sCR} \right)} \right)^2 \quad R = \frac{b}{\sqrt{4\eta^4 + b^2 C^2 \omega_0^2}}$$

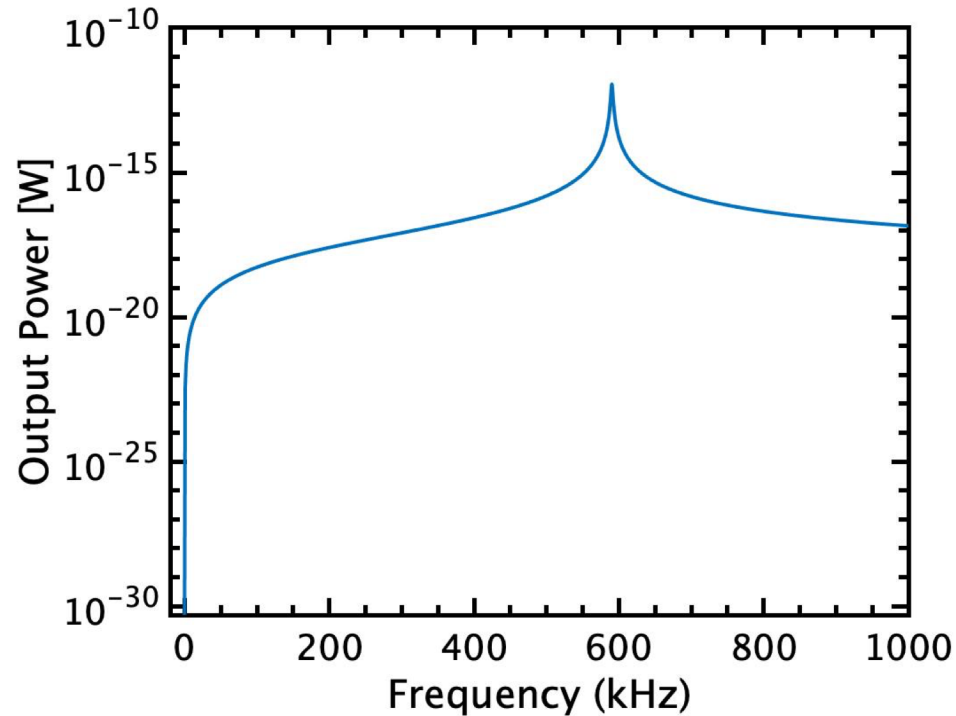
As η increases,

$$R \rightarrow \frac{b}{2\eta^2}, \quad \frac{1}{1 + sCR} \rightarrow 1, \quad P_{resonance} \rightarrow \frac{f^2}{8b}$$

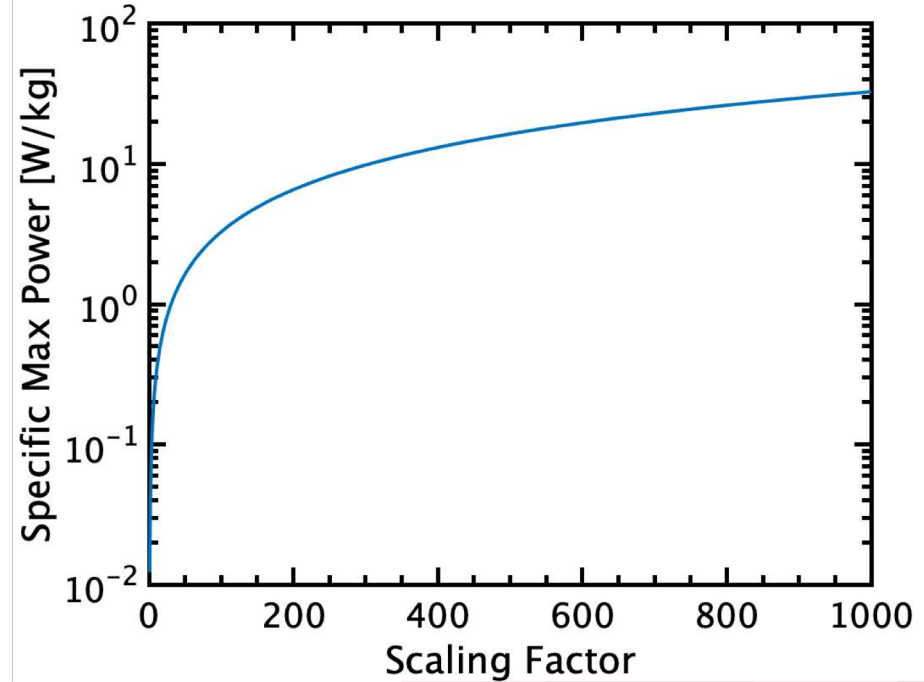
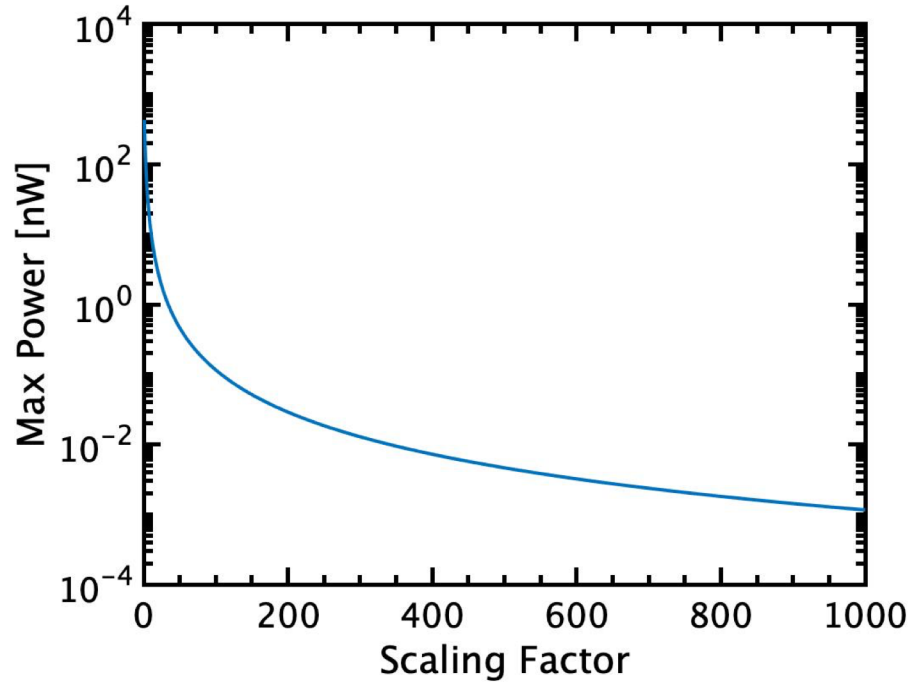
Physical interpretation:

- The resistive load dominates the output impedance
- The increased coupling decreases optimal load and counteracts increase in current

Scaling the Entire Structure 1000x



Trend of Scaling the Entire Structure



Final Structure Specifications

Description	Original	Gap Scaled (S=7.5)	Fully Scaled (S=1000)
Mass	35.25 mg	35.25 mg	35.25 pg
Thickness	300 μm	300 μm	300 nm
Beam Length	$\sim 1400 \mu\text{m}$	$\sim 1400 \mu\text{m}$	1.4 μm
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Conclusion

	Original	Gap Scaled (S=7.5)	Fully Scaled (S=1000)
Power output (per mass)	0.16 W/kg	0.74 W/kg	7.37 kW/kg
Resonant Frequency	590 Hz	590 Hz	590 kHz
Bandwidth (FWHM)	1.28 Hz	2.28 Hz	2284.6 Hz
Peak specific power output (per mass)	0.0122 W/kg	0.033 W/kg	32.78 W/kg
Load Resistance	17.5 M Ω	0.19 M Ω	76.4 k Ω

Scaling down the device:

- Increases output power per mass
- Raises resonance frequency
 - Problem for spectrums with low frequency
- Decreases needed bias voltage
- Drives low load resistance
- Increases bandwidth

References

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- [4] Nguyen, D. S., et al. "Fabrication and characterization of a wideband MEMS energy harvester utilizing nonlinear springs." *Journal of Micromechanics and Microengineering* 20.12 (2010): 125009.
- [5] Cepnik, C and Wallrabe, U. 2011. A Micro Energy Harvester with 3D Wire Bonded Microcoils. (Beijing, China: IEEE, Transducers'11) pp 665-8
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